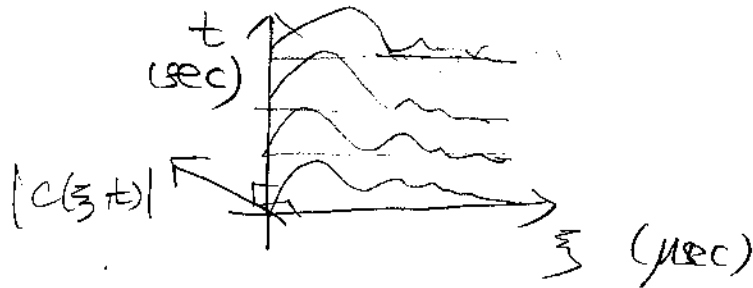


## Comments on Deterministic $C(\xi, t)$

①  $C(\xi, t) = h_t(\xi) =$  impulse response at  $t$ ,  
with relative delay  $\xi$

② Usually, time scale in  $\xi$  is much smaller  
than time scale in  $t$



$\Rightarrow$  We may consider Fourier transform of  $C(\xi, t) = h_t(\xi)$  with respect to  $\xi$ .

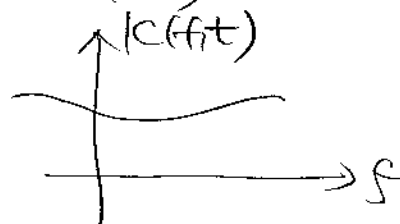
Then,

$$C(f, t) = H_t(f) \triangleq \int_{-\infty}^{\infty} C(\xi, t) e^{j2\pi f \xi} d\xi$$

represents the frequency response of  $C(\xi, t)$  at  $t$ .

•  $C(f, t)$

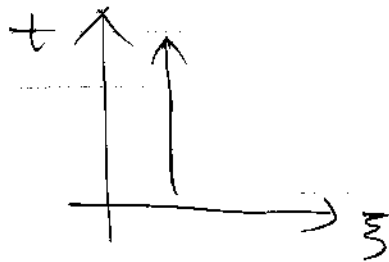
If  $|C(f, t)|$  varies very slowly in  $f$  given fixed  $t$ ,  
then, frequency coherence is very high



Let coherence bandwidth be the <sup>maximum</sup> distance b/w  
two  $f$ 's having almost the same value of  $|C(f, t)|$ .

As we've seen already, the coherence bandwidth times the maximum delay spread determines the severity of frequency selectivity.

② Now consider  $c(\xi, t)$  as a function of  $t$



$$\text{ex/ } c(\xi, t) = x(t) \delta(\xi - \Delta)$$

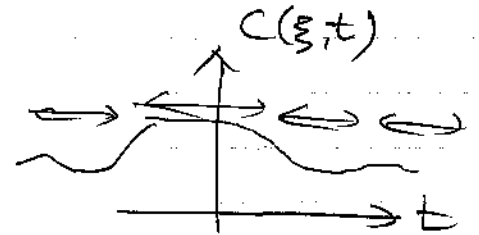
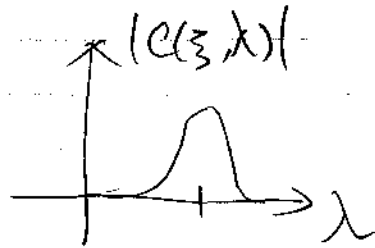
$$\Rightarrow y(t) = x(t) x(t - \Delta)$$

$$\Rightarrow Y(f) = X(f) * X(f) e^{-j2\pi f \Delta}$$

$\Rightarrow$  From  $c(\xi, t)$  as a function of  $t$ , we can get the information about how fast the channel changes.

Consider Fourier transform of  $c(\xi, t)$  with respect to  $t$ .

$$\int_{-\infty}^{\infty} c(\xi, t) e^{j2\pi f t} dt = C(\xi, \lambda)$$



We can define coherence time from  $c(\xi, t)$  by finding the time interval given fixed  $\xi$ , the length of

If  $|C(\xi, \lambda)|$  has energy concentration in  $\lambda_0$  then  $\lambda_0$  represents the Doppler frequency.

# Wireless Channels : WSSUS channel model

## ○ Complex baseband representation

• From the relation

$$y(t) = \int_{-\infty}^{\infty} c(\xi, t) x(t-\xi) d\xi$$

we define

$$c_c(\xi, t) = c(\xi, t) e^{-j2\pi f_c \xi}$$

where  $c(\xi, t)$  is bandpass for each  $t$ .

$$(\therefore c(\xi, t) = \text{Re} \{ c_c(\xi, t) e^{j2\pi f_c \xi} \})$$

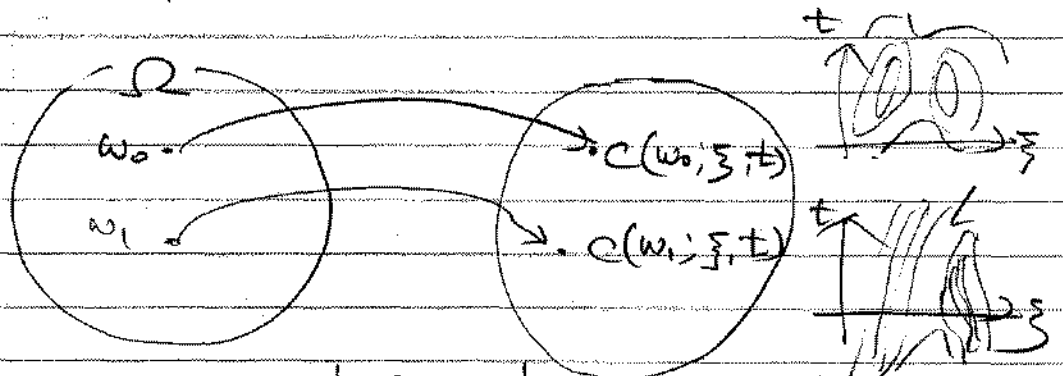
Then,

$$y(t) = \int_{-\infty}^{\infty} \frac{1}{2} c_c(\xi, t) x(t-\xi) d\xi$$

• In this note, we use  $c(\xi, t)$  to denote  $\frac{1}{2} c_c(\xi, t)$ .

## ○ Proper-complex Gaussian approximation.

• In some applications, we can model  $c(\xi, t)$  as a proper-complex Gaussian random field.



So, the envelope  $|c(\xi_0, t_0)|$  is modeled as

a Rayleigh random variable if  $E\{c(\xi, t)\} = 0$   
while

a Rician " " " " " "  $\neq 0$ .

### ○ Channel correlation property 1 (Time-Domain) proper-complex

- Since  $c(\xi, t_0)$  is modeled as a Gaussian random variable and  $c(\xi, t)$  for each  $(\xi, t)$ -pair are modeled as joint Gaussian, the statistical property of the random field  $c(\xi, t)$  is completely characterized by

$$(i) E\{c(\xi, t)\} = 0 \quad (*)$$

$$(ii) E\{c(\xi_0, t_0)^* c(\xi_1, t_1)\} \text{ for every } \xi_0, \xi_1, t_0, t_1$$

We assume zero-mean in what follows: ... (\*\*)

- A common assumption 1 is that

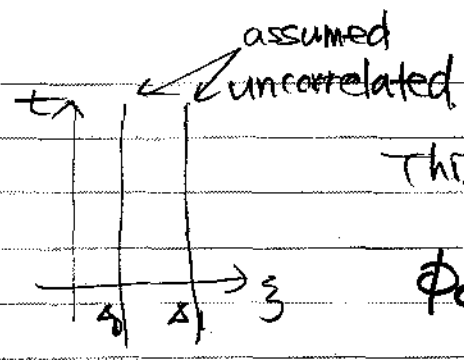
random processes  $c(\xi, t)$  are assumed jointly WSS for any choice of  $\xi_0$  and  $\xi_1$ . (WSS model)

This with the Gaussian assumption leads (\*\*) to the following definition.

$$\phi_c(\xi_1, \xi_2; \tau) \triangleq E\{c(\xi_1, t)^* c(\xi_2, t+\tau)\} \quad (***)$$

- A common assumption 2 is that

$c(\xi_0, t)$  and  $c(\xi_1, t)$  are uncorrelated random processes. (Uncorrelated Scattering model)



This leads (let's) to

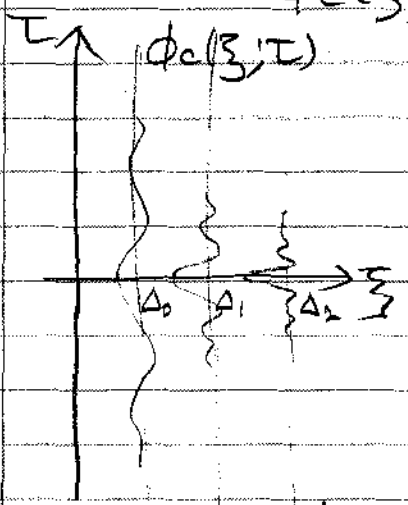
$$\phi_c(\xi_1, \xi_2; \tau) = \phi_c(\xi_1; \tau) \delta(\xi_1 - \xi_2)$$

In summary, WSSUS model leads to

$$E\{C(\xi_1, t_1)^* C(\xi_2, t_2)\} = \phi_c(\xi_1; t_1 - t_2) \delta(\xi_1 - \xi_2)$$

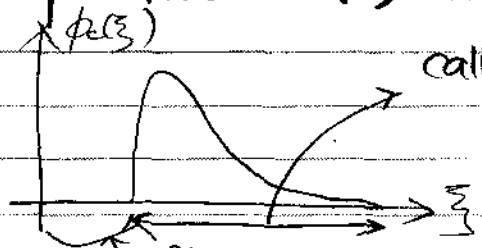
which shows the necessity to define the auto-correlation function

$$\phi_c(\xi; \tau) \triangleq E\{C(\xi, t)^* C(\xi, t + \tau)\}$$



- Note that  $\phi_c(\xi; 0)$  denotes the power of each uncorrelated random process at delay  $\xi$ . We call  $\phi_c(\xi) \triangleq \phi_c(\xi; 0)$  the multipath

intensity profile (MIP) or Power delay profile



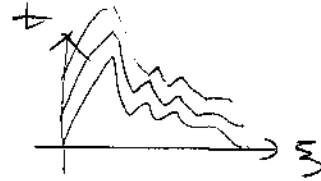
called multipath/delay spread of the channel.

we often ignore this channel delay

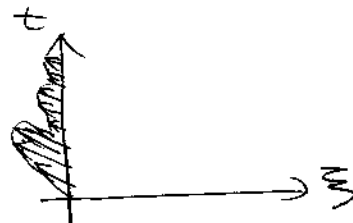
# Two Extreme Cases for WSSUS channel

○ Two extreme cases

(i)  $C(\xi, t) = h(\xi)$



(ii)  $C(\xi, t) = \alpha(t) \delta(\xi)$



○ Random static channel

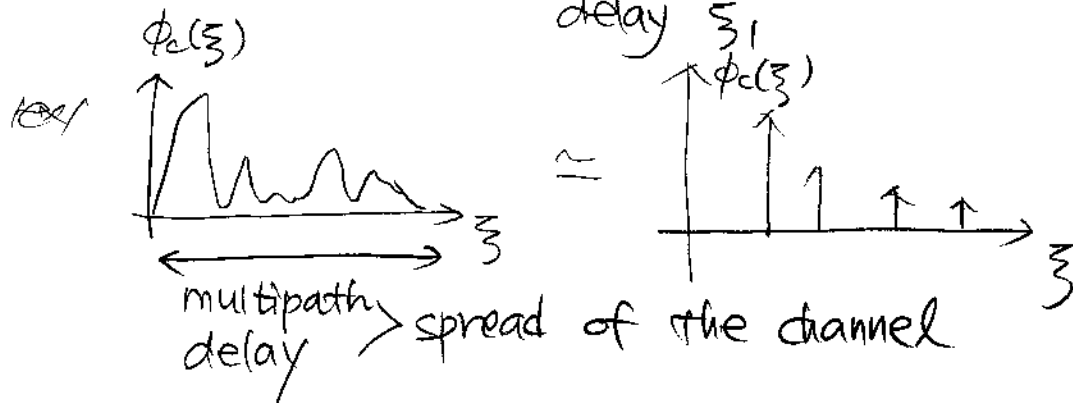
•  $C(\xi, t) = h(\xi)$

•  $E\{h(\xi)\} = 0, \forall \xi$   
 •  $E\{h(\xi_1) h^*(\xi_2)\} = \phi_c(\xi_1) \delta(\xi_1 - \xi_2) \quad \because \text{uncorrelated}$

$\underbrace{\phi_c(\xi_1)}_{\triangleq E\{|h(\xi_1)|^2\}}$  multipath intensity profile (MIP)

= variance of channel response delay  $\xi_1$

= power of the component w/ delay  $\xi_1$



$$E\{H(f)\} = E\left\{ \int_{-\infty}^{\infty} h(\zeta) e^{-j2\pi f \zeta} d\zeta \right\} = 0, \forall f$$

$$E\{H(f_1)^* H(f_2)\} = E\left\{ \left( \int_{-\infty}^{\infty} h(\zeta_1) e^{-j2\pi f_1 \zeta_1} d\zeta_1 \right)^* \left( \int_{-\infty}^{\infty} h(\zeta_2) e^{-j2\pi f_2 \zeta_2} d\zeta_2 \right) \right\}$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \underbrace{E\{h(\zeta_1)^* h(\zeta_2)\}}_{\phi_c(\zeta_1) \delta(\zeta_1 - \zeta_2)} e^{j2\pi(f_1 \zeta_1 - f_2 \zeta_2)} d\zeta_1 d\zeta_2$$

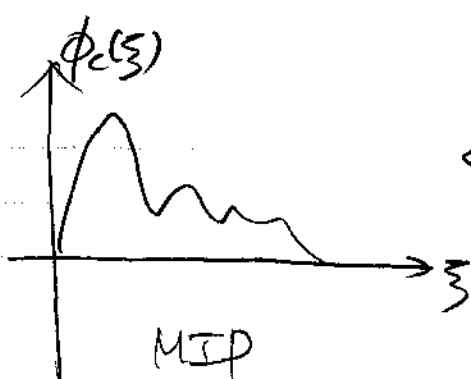
$$= \int_{-\infty}^{\infty} \phi_c(\zeta_1) e^{j2\pi(f_1 - f_2)\zeta_1} d\zeta_1$$

$$\cong \phi_c(f_1 - f_2)$$

Fourier transform

of the MIP evaluated at  $f_2 - f_1$

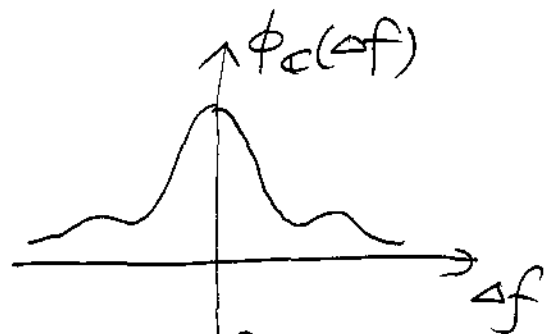
"Spaced frequency coherence function"



MIP

↓  
delay spread

$\mathcal{F}$



Spaced frequency correlation function

↓  
coherence bandwidth

is determined by the MIP.

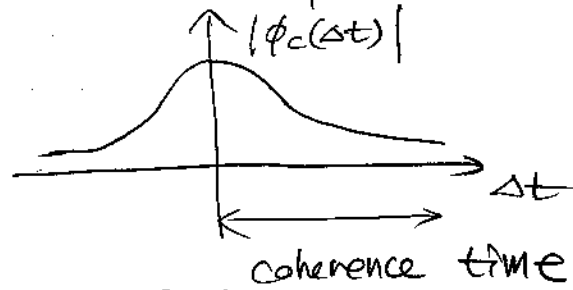
roughly inversely proportional

If coherence BW  $\gg$  signal BW, then frequency flat. Otherwise, "selective."

### 0 Random Multiplicative Channel

- $\alpha(\xi, t) = \alpha(t) \delta(\xi)$
- $E\{\alpha(t)\} = 0, \forall t$
- $E\{\alpha(t)^* \alpha(t+\tau)\} = \phi_c(\tau), \quad \text{WSS}$
- or  $E\{\alpha(t_1)^* \alpha(t_2)\} = \phi_c(t_2 - t_1) = \phi_c(\Delta t)$

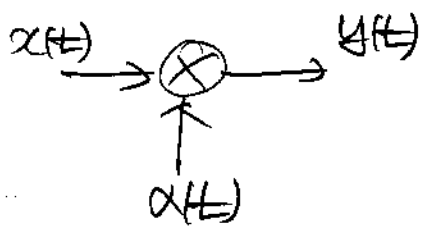
spaced time correlation function.



autocorrelation function of the multiplicative process.

FD autocorrelation  $\longleftrightarrow$  power spectrum

$$S_c(\lambda) = \mathcal{F}\{\phi_c(\Delta t)\} = \int_{-\infty}^{\infty} \phi_c(\Delta t) e^{-j2\pi\lambda\Delta t} d\Delta t$$



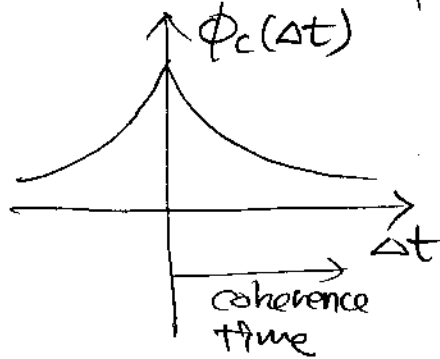
$$\begin{aligned} S_{yy}(f) &= \mathcal{F}\{A[E\{y(t_1)^* y(t_2)\}]\} \\ &= \mathcal{F}\{A[\phi_c(t_2 - t_1) R_{xx}(t_2 - t_1)]\} \\ &= \mathcal{F}\{\phi_c(t_2 - t_1) \tilde{R}_{xx}(t_2 - t_1)\} \\ &= \mathcal{F}\{\phi_c(\Delta t)\} * S_{xx}(f) \end{aligned}$$

approximate using deltas



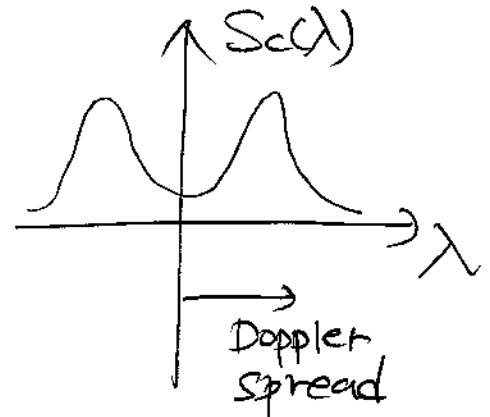
$$S_{xx}(\lambda) \rightarrow \boxed{S_c(\lambda)} \rightarrow S_{FF}(\lambda)$$

$S_{xx}(\lambda)$  is convolved w/  $S_c(\lambda)$   
 $S_{xx}(\lambda)$  is spread by  $S_c(\lambda)$



spaced-time correlation function

$\mathcal{F}$



Doppler power spectrum

○ Channel correlation property 2 (Freq. - Domain)

zero-mean proper-complex

- For WSSUS channel model,

$$\phi_c(\xi, \Delta t) \triangleq E\{C(\xi, t)^* C(\xi, t + \Delta t)\}$$

completely characterizes the channel, because

- $\phi_c(\xi_1, \xi_2; \Delta t) = \phi_c(\xi_1, \Delta t) \delta(\xi_1 - \xi_2)$ .

- Now consider a 2-D Fourier transform of  $\phi_c(\xi_1, \xi_2; \Delta t)$  as

$$\begin{aligned} \phi_c(f_1, f_2; \Delta t) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \phi_c(\xi_1, \xi_2; \Delta t) e^{j2\pi(f_1 \xi_1 - f_2 \xi_2)} d\xi_1 d\xi_2 \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \phi_c(\xi_1; \Delta t) \delta(\xi_1 - \xi_2) e^{j2\pi(f_1 \xi_1 - f_2 \xi_2)} d\xi_1 d\xi_2 \\ &= \int_{-\infty}^{\infty} \phi_c(\xi_1; \Delta t) e^{j2\pi(f_2 - f_1) \xi_1} d\xi_1 \\ &\triangleq \phi_c(f_2 - f_1; \Delta t) \end{aligned}$$

This shows

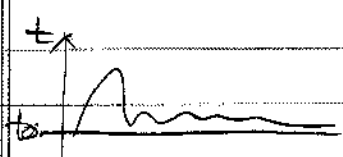
$\phi_c(f_2 - f_1; t_2 - t_1)$  completely characterizes the channel.

called the spaced-frequency, spaced-time correlation function.

- Note that

$$\phi_c(f_1, f_2; \Delta t) = E\left[ \left( \int_{-\infty}^{\infty} C(\xi_1, t) e^{-j2\pi f_1 \xi_1} d\xi_1 \right)^* \left( \int_{-\infty}^{\infty} C(\xi_2, t + \Delta t) e^{-j2\pi f_2 \xi_2} d\xi_2 \right) \right]$$

Since  $\int_{-\infty}^{\infty} c(\xi, t_0) e^{-j2\pi f \xi} d\xi \triangleq C(f, t_0)$



and  $Y(f) = \int_{-\infty}^{\infty} c(\xi, t_0) x(t_0 - \xi) d\xi$

$\xrightarrow{\xi} = c(\xi, t_0) * x(\xi) \Big|_{\xi=t_0} = \int_{-\infty}^{\infty} C(f, t_0) X(f) e^{j2\pi f t_0} df$

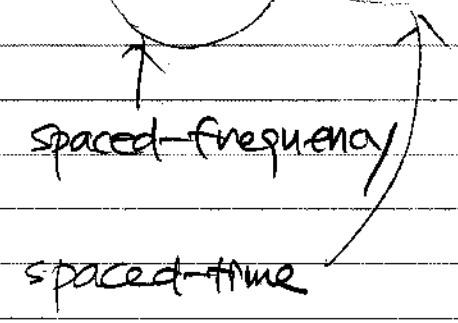
$C(f, t_0)$  means the weighting of the signal  $X(f)$  at  $f=f_0$ .

As time changes, the weighting at  $f_0$  changes.

Therefore,

$\phi_c(\Delta f, \Delta t)$

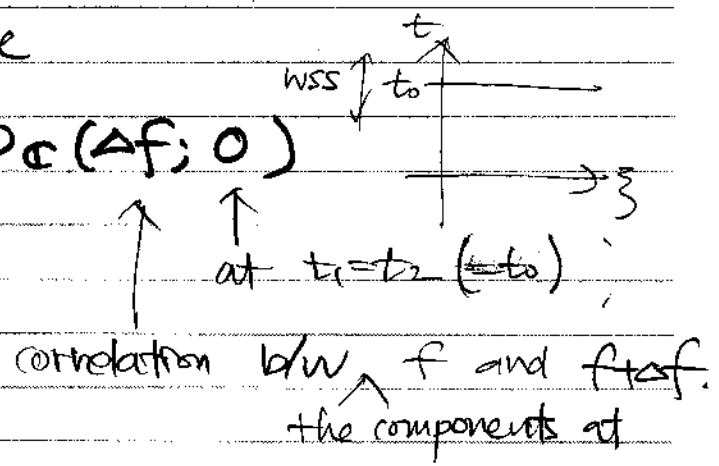
means the correlation b/w  $C(f, t)$  and  $C(f + \Delta f, t + \Delta t)$ .



Consider  $\tau=0$  and define

$\phi_c(\Delta f) \triangleq \phi_c(\Delta f; 0)$

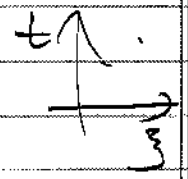
tells us frequency coherence of the channel.



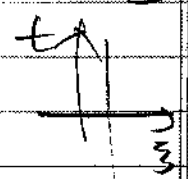


# Channel correlation property 3 (frequency Domain)

Review of what we have done so far.

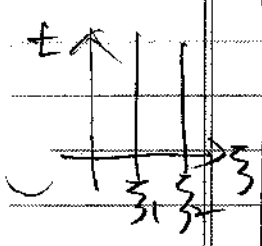


$C(\xi, t)$  modeled as a zero-mean proper-complex Gaussian random field.



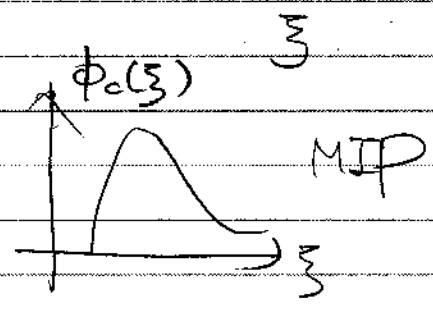
$C(\xi_0, t)$ 's are modeled as jointly WSS uncorrelated random processes

$$E[C(\xi_1, t_1)^* C(\xi_2, t_2)] = \phi_c(\xi_1, t_2 - t_1) \delta(\xi_1 - \xi_2)$$

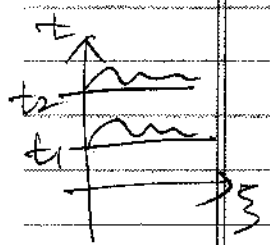


$\phi_c(\xi, \tau)$  characterizes the property auto-correlation function at  $\xi$ .

$\phi_c(\xi, 0)$  tells us the power at each



space-frequency  
space-time  
correlation function.



$$E[C(f_1, t_1)^* C(f_2, t_2)] = \phi_c(f_1, f_2; t_2 - t_1)$$

$$\hat{=} \phi_c(\Delta f; \Delta t)$$

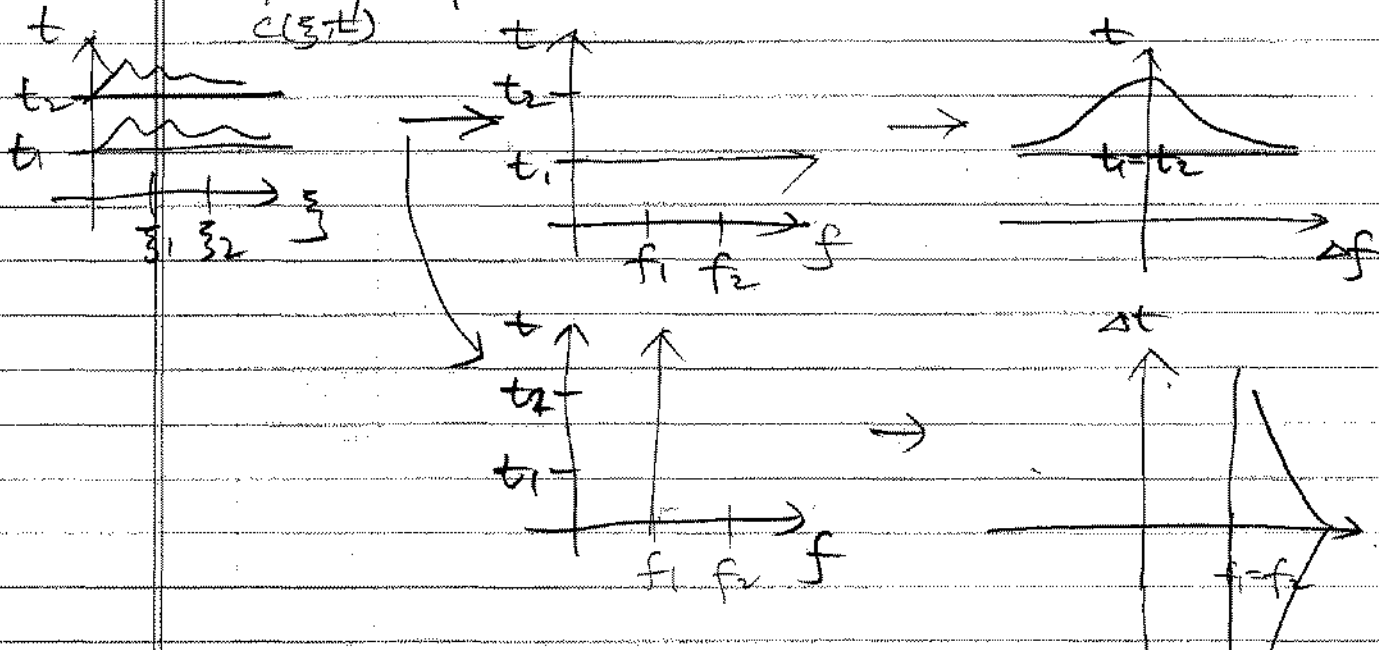
$$t_1 = t_2 \Rightarrow \mathcal{F}_{\xi} \{ \phi_c(\xi, \Delta t) \}$$

$$\phi_c(\Delta f; 0) = \mathcal{F}_{\xi} \{ \phi_c(\xi) \}$$

coherence  
bandwidth

MIP  
delay  
spread

- Now, we focus on  $f_1 = f_2$  case of the spaced-frequency spaced-time correlation function



Define

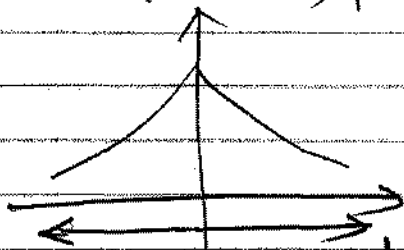
$$S_c(\Delta f; \lambda) = \int_{-\infty}^{\infty} \phi_c(\Delta f; \Delta t) e^{j2\pi\lambda \Delta t} d\Delta t$$

Set  $\Delta f = 0$  and define

$$S_c(\lambda) \equiv S_c(0, \lambda)$$

$$|\phi_c(0; \Delta t)| = \int_{-\infty}^{\infty} \phi_c(0; \Delta t) e^{j2\pi\lambda \Delta t} d\Delta t$$

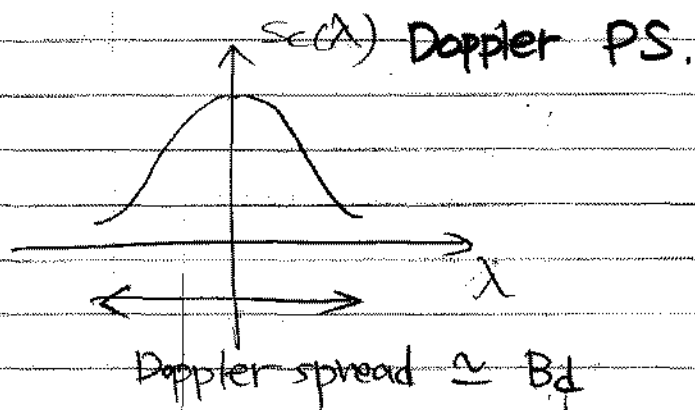
autocorrelation of a frequency component at different time instants  $t$  &  $t + \Delta t$



coherence time  $\approx \frac{1}{B_d}$

Spaced-time autocorrelation function

$S_c(\lambda)$  is called the Doppler power spectrum of the channel.



• Why called Doppler spread?

Let/  $c(\xi, t) = \alpha(t) \delta(\xi)$

$$\phi_c(f_1, f_2; \Delta t) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} E [c(\xi_1, t)^* c(\xi_2, t + \Delta t)] e^{j2\pi(f_1 \xi_1 - f_2 \xi_2)} d\xi_1 d\xi_2$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} E [\alpha(t)^* \alpha(t + \Delta t)] \delta(\xi_1 - \xi_2) e^{j2\pi(f_1 \xi_1 - f_2 \xi_2)} d\xi_1 d\xi_2$$

$$= \int_{-\infty}^{\infty} E [\alpha(t)^* \alpha(t + \Delta t)] e^{j2\pi \Delta f \xi_1} d\xi_1$$

$$= E [\alpha(t)^* \alpha(t + \Delta t)] \delta(\Delta f)$$

$$\underbrace{\hspace{10em}}_{\phi_c(\Delta t)}$$

$$S_c(\lambda) = \mathcal{FT} \{ \phi_c(\Delta t) \}$$

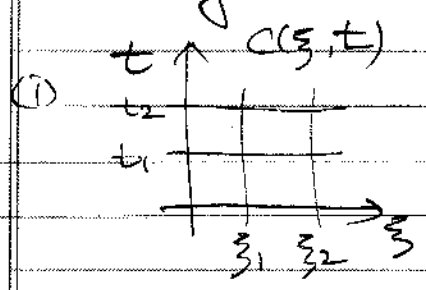
○ Scattering function.

2-D Fourier transform of space-time correlation function  $\phi_c(\Delta f; \Delta t)$

$$S(\tau; \lambda) \triangleq \mathcal{DFT} \{ \phi_c(\Delta f; \Delta t) \}$$

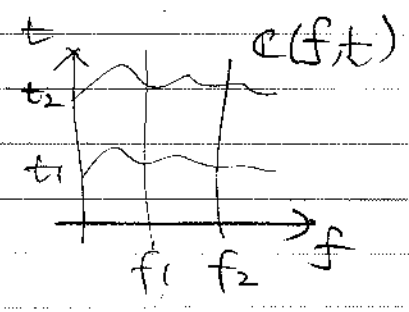
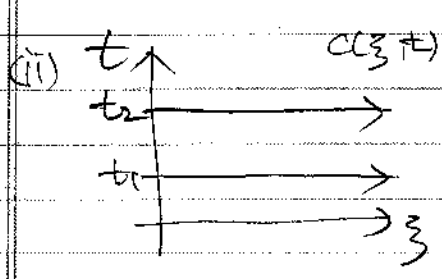
Summary

space-time correlation function



$$E [C(\xi_1, t_1)^* C(\xi_2, t_2)] = \phi_C(\xi_1, \underbrace{t_2 - t_1}_{\Delta t}) \delta(\xi_1 - \xi_2)$$

∴ WSSUS.



$$E [C(f_1, t_1)^* C(f_2, t_2)] \triangleq \text{DF} \{ \nu \}$$

spaced-freq  
spaced-time  
correlation ft.  $\rightarrow$   $\phi_C(\Delta f, \Delta t)$

∴ WSSUS.

(iii)  $\phi_C(\Delta f, 0)$  spaced-freq. correlation ft.  
 $\rightarrow$  coherence bandwidth  
 $\uparrow \Delta f$

$\phi_C(\Delta \xi)$  Multipath Intensity profile (MIP)  
 $\rightarrow$  delay spread

(iv)  $\phi_C(0, \Delta t)$  spaced-time correlation ft.  
 $\rightarrow$  coherence time  
 $\uparrow \Delta t$

$S_C(\lambda)$  Doppler power spectrum  
 $\rightarrow$  doppler spread

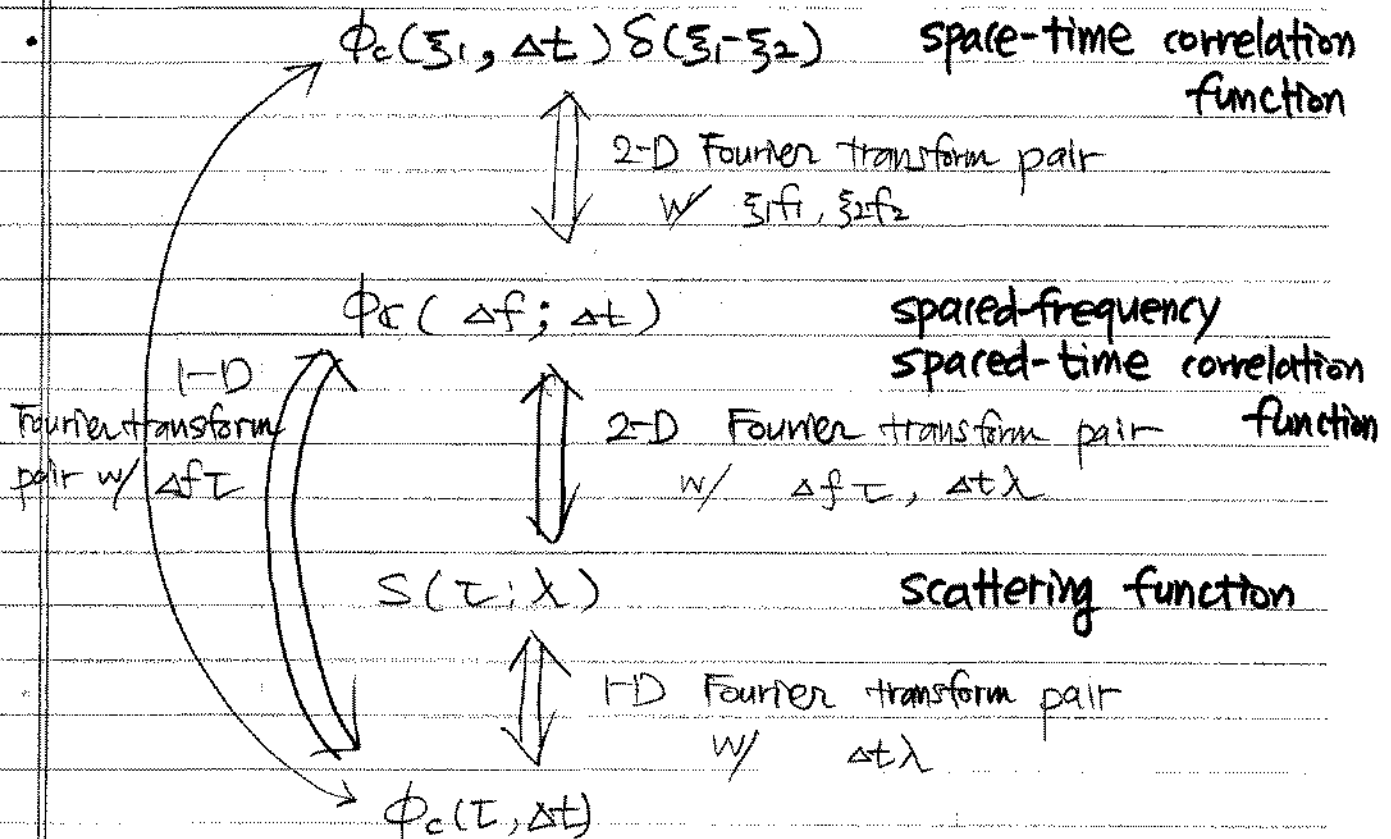


# Wireless Channels : Scattering function

## ○ Doubly spread channel (PZB 8-2.3)

- A doubly spread channel is a WSSUS channel that spreads the signal transmitted through the channel in time and frequency.
- A singly spread degenerate channel is a WSSUS channel that spreads the signal transmitted through the channel only in one domain.

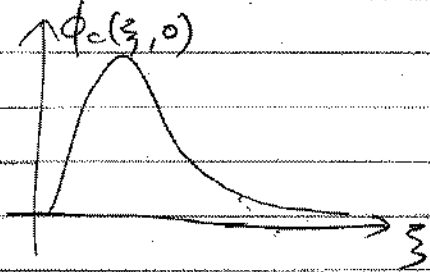
## ○ Relation b/w correlation functions



## ○ Scattering function

(i) The MIP was defined as  $\phi_c(\xi, 0)$  and the

delay spread was defined as the length of the interval that contains the support of  $\phi_c(\xi, 0)$ .



The related definitions are the mean delay and the mean-square delay spread defined as

$$M_R = \frac{\int_{-\infty}^{\infty} \xi \phi_c(\xi, 0) d\xi}{\int_{-\infty}^{\infty} \phi_c(\xi, 0) d\xi}$$

and

$$L = \frac{\int_{-\infty}^{\infty} \xi^2 \phi_c(\xi, 0) d\xi}{\int_{-\infty}^{\infty} \phi_c(\xi, 0) d\xi} - M_R^2$$

respectively.

In terms of the scattering functions,

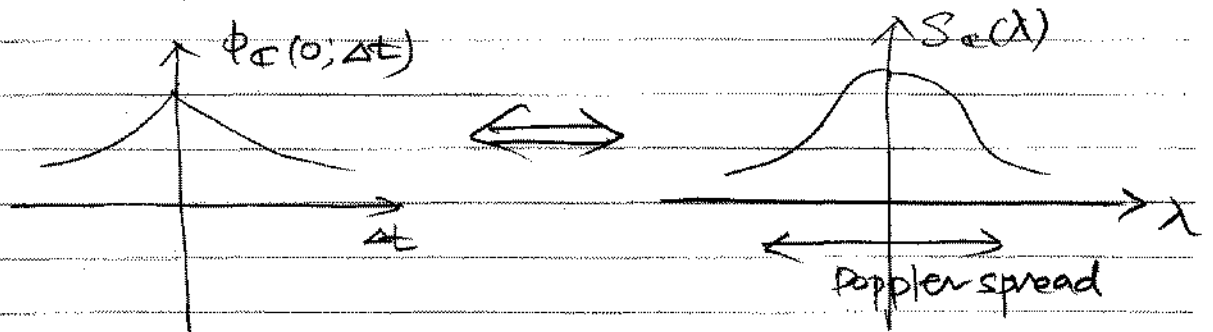
$$M_R = \frac{\int_{-\infty}^{\infty} \xi \int_{-\infty}^{\infty} S(\xi, \lambda) e^{j2\pi \Delta t \lambda} d\lambda \Big|_{\Delta t=0} d\xi}{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S(\tau, \lambda) d\lambda d\tau}$$

$$= \frac{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \tau S(\tau, \lambda) d\lambda d\tau}{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S(\tau, \lambda) d\lambda d\tau}$$

and

$$L = \frac{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \tau^2 S(\tau, \lambda) d\lambda d\tau}{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S(\tau, \lambda) d\lambda d\tau} - M_R^2$$

(ii) The Doppler spread was defined in terms of  $S_c(x)$ , which is the Fourier transform of  $\phi_c(0, \Delta t)$ .



The related definitions are the mean Doppler spread and the mean-square Doppler spread defined as

$$M_R = \frac{\int_{-\infty}^{\infty} \lambda S_c(x) d\lambda}{\int_{-\infty}^{\infty} S_c(x) d\lambda}$$

$$= \frac{\int_{-\infty}^{\infty} \lambda \int_{-\infty}^{\infty} \phi_c(\Delta f, \Delta t) e^{j2\pi\lambda\Delta t} d\Delta t \Big|_{\Delta f=0} d\lambda}{\dots}$$

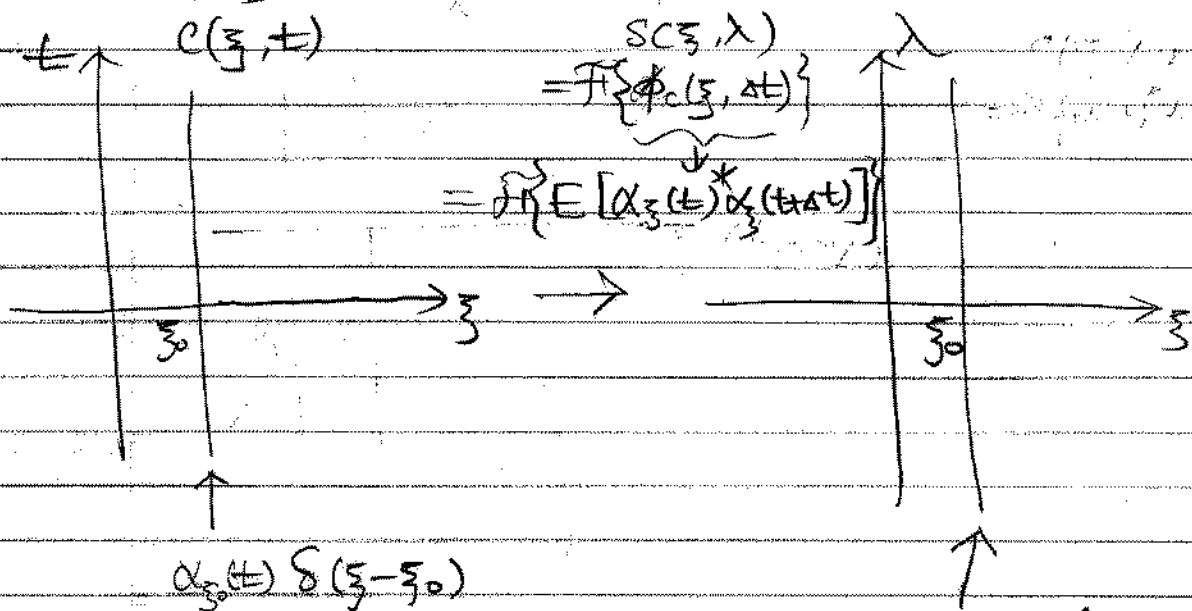
$$= \frac{\int_{-\infty}^{\infty} \lambda \int_{-\infty}^{\infty} S(\tau, \lambda) e^{j2\pi\lambda\tau} d\tau \Big|_{\Delta f=0} d\lambda}{\dots}$$

$$= \frac{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \lambda S(\tau, \lambda) d\tau d\lambda}{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S(\tau, \lambda) d\tau d\lambda}$$

$$B = \frac{\int_{-\infty}^{\infty} \lambda^2 S_c(x) d\lambda}{\int_{-\infty}^{\infty} S_c(x) d\lambda} \quad MR^2$$

$$= \frac{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \lambda^2 S(\tau, \lambda) d\lambda d\tau}{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S(\tau, \lambda) d\lambda d\tau} \quad MR^2$$

- Thus, the scattering function  $S(\xi, \lambda)$  gives us the information about the relation b/w the propagation delay and associated multiplicative process PSD.



contains the information of the power spectrum of  $\alpha_{\xi_0}(t)$ .

Since  $\alpha_{\xi_0}(t)$  is strictly stationary,  $\alpha_{\xi_0}(t)\delta(t-\xi_0)$  has the PSD that is the convolution of  $S(\xi_0, \lambda)$  and the PSD of  $S(t-\xi_0)$ .