

Zero-Forcing Decision Feedback Equalizer (ZF-DFE) ①

○ LE vs DFE

So far, we learned about two linear equalizers:

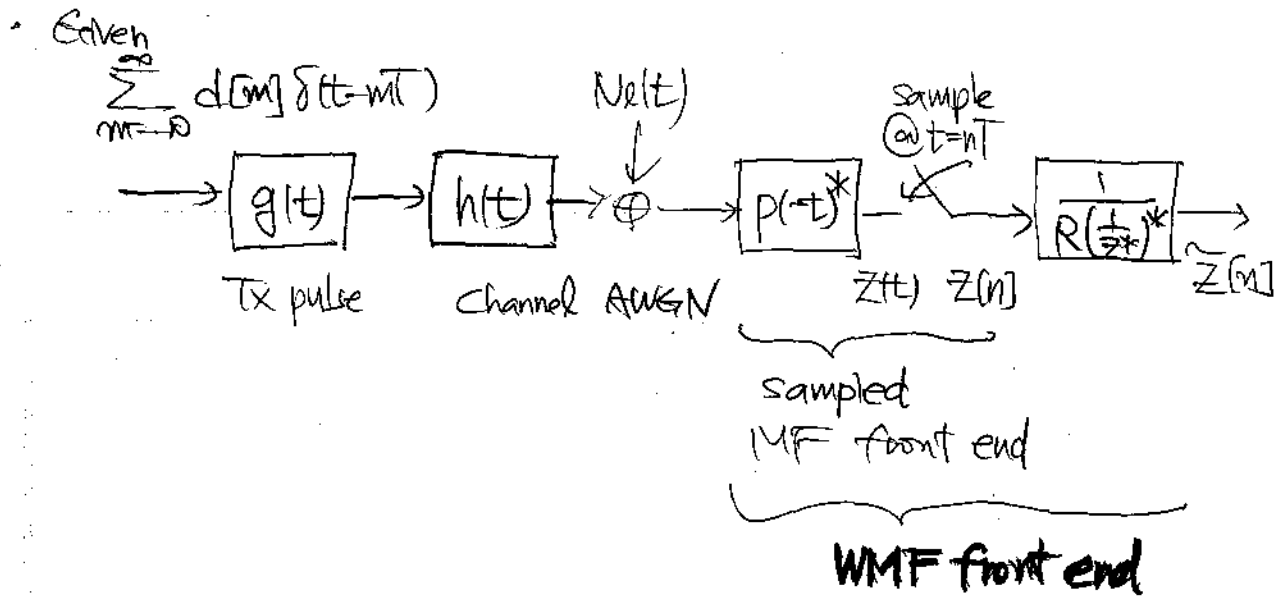
- ◀ ZF-LE and
- ◀ MMSE-LE.

In this handout, we learn ^{about} a non-linear equalizer. As most of the non-linear system designs, we slightly modify a well-known linear system by inserting a very simple non-linear block.

There are two approaches in obtaining the ZF-DFE:

- ◀ WMF front-end
- ◀ ZF-LE

○ Derivation of ZF-DFE from WMF front end



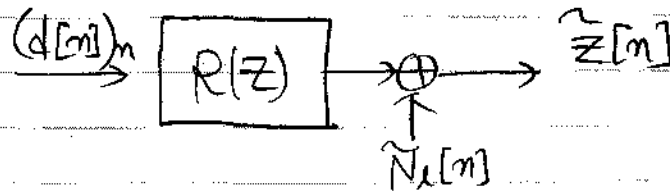
$$\hat{z}[n] = \sqrt{2P} \sum_{m=-\infty}^{\infty} d[m] r[n-m] + \tilde{N}_e[n]$$

where $\beta(mT) \stackrel{z}{\leftrightarrow} Q(z) = R(z)R\left(\frac{1}{z^*}\right)^*$

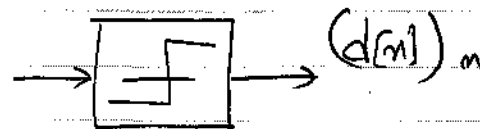
with $E\{\tilde{N}_e[m]^* \tilde{N}_e[m+n]\} = 2N_0 \delta[m]$

AWGN sequence

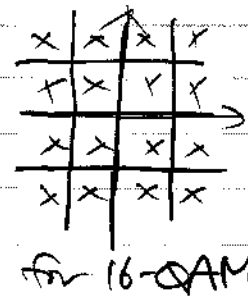
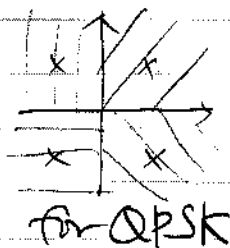
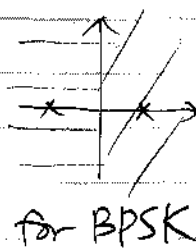
- The end-to-end response can be represented by



- The key assumption in the use of a slicer is that



the slicer output is perfect.



- The objective in the design of a DFE is to make the slicer input as "good" as possible, so that the slicer output is close to $(d[m])_m$.

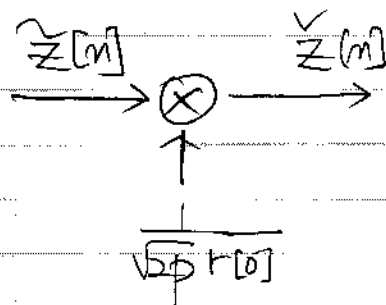
- In ZF-DFE, the goodness is defined as

(i) The slicer input contains no ISI but $d[m]$ only as the signal component. \rightarrow ZF

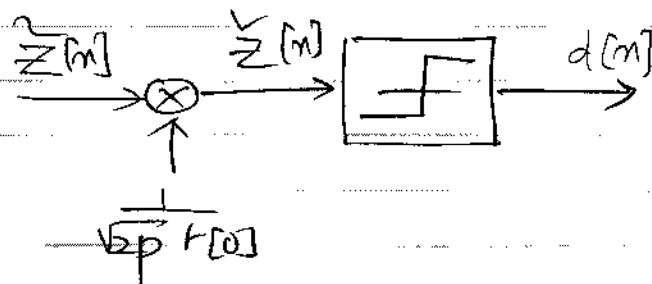
(ii) The slicer input SNR is maximized

(iii) The slicer input noise sequence is white.

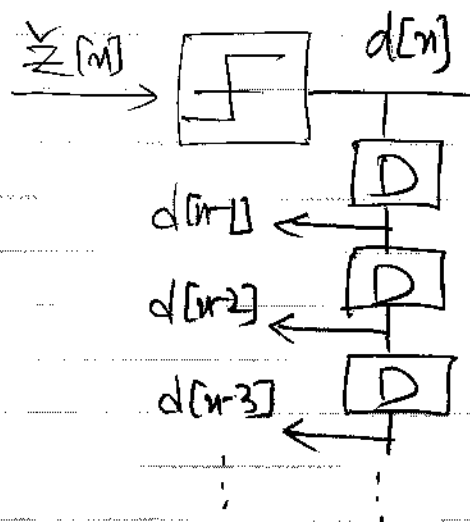
- If we directly feed the slicer with $\tilde{z}[n]$, then (i) is violated. In particular, $d[n]$ even has the coefficient $\sqrt{2p}r[0]$. So, we first scale $\tilde{z}[n]$ as



- If we feed the slicer with $\check{z}[n]$, then



the slicer output is assumed to be $d[n]$.
By delaying this output we can obtain $d[n-1], d[n-2], d[n-3], \dots$ as



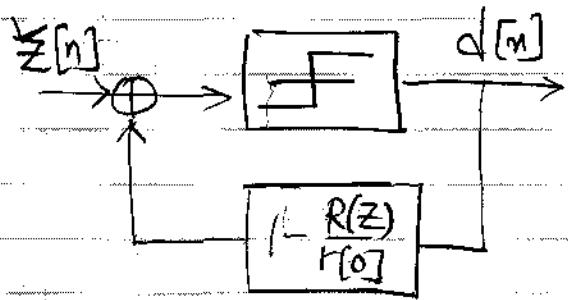
To meet condition (i), we use these past decisions as

$$\hat{z}[n] = \frac{r[1]}{r[0]} d[n-1] + \frac{r[2]}{r[0]} d[n-2] + \dots$$

$$= d[n] + \frac{r[1]}{r[0]} d[n-1] + \dots$$

However, the precursors remain if $r[n]$ is not causal. So, we need **causal $r[n]$**

Then,



perfectly cancels the ISI.

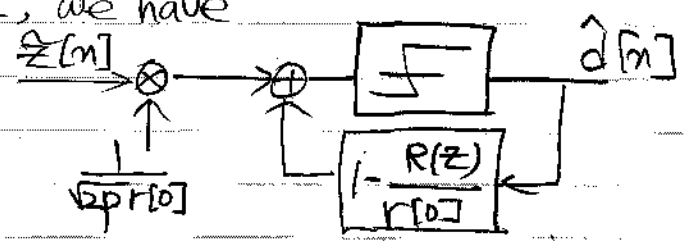
Now, take a look at the noise part. The slicer input is now given by

$$d[n] + \frac{\hat{N}_e[n]}{\sqrt{2P} |r[0]|}$$

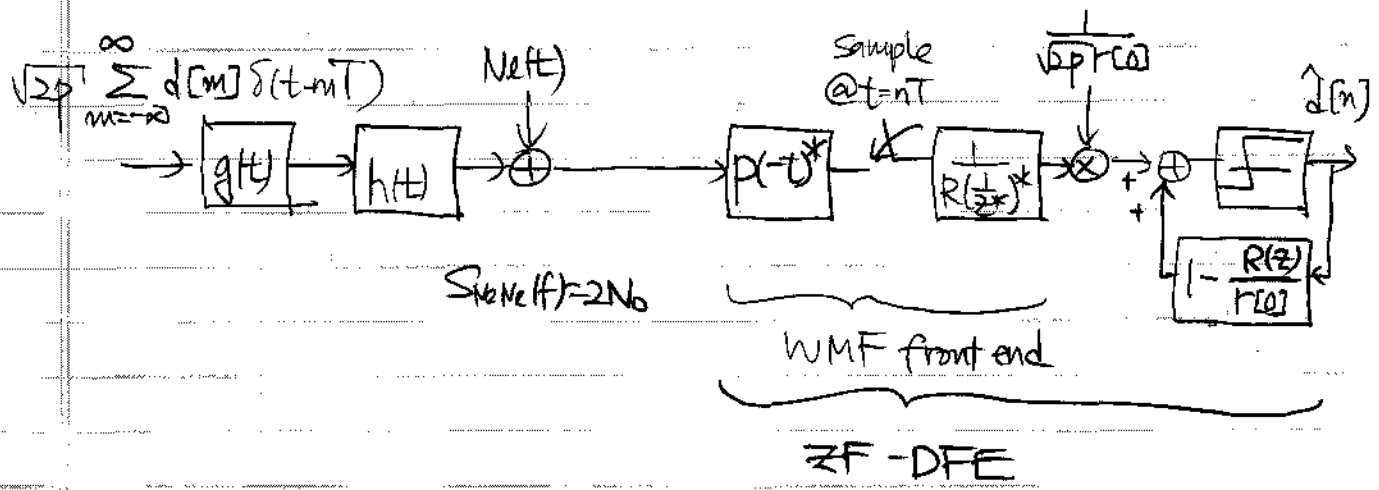
where $(\hat{N}_e[n])_n$ is **white** with variance $\frac{2N_0}{2P|r[0]|^2}$

So, the input SNR is maximized by choosing the minimum phase $R(z)$.

Therefore, we have



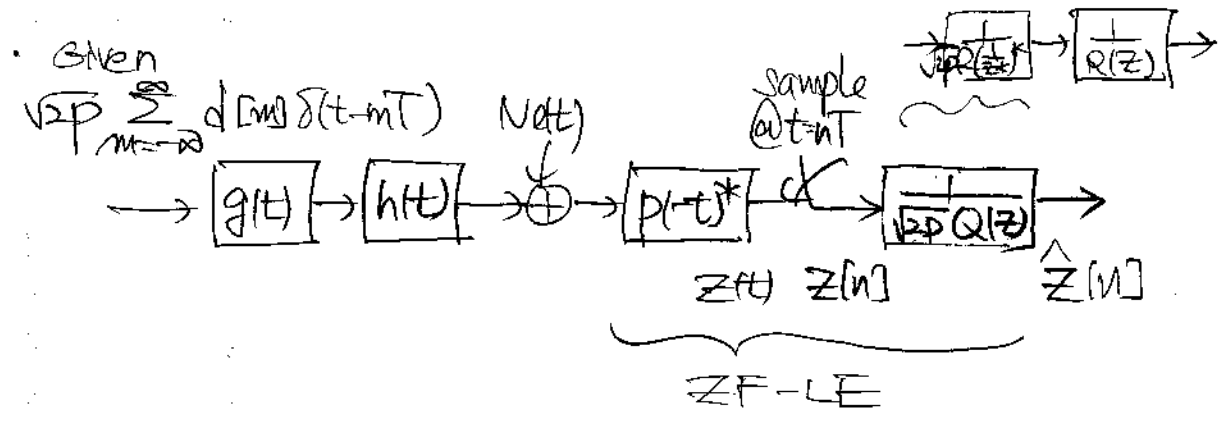
Combined w/ the WMF front-end, the ZF-DFE is given by



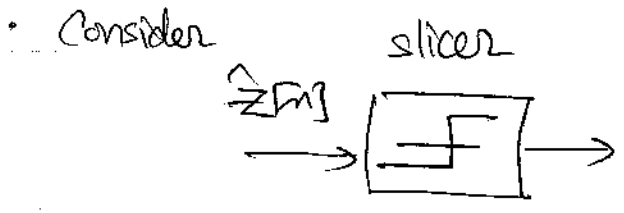
where $R(z)$ must be the minimum phase system that factors $Q(z)$ as $Q(z) = R(z)R(\frac{1}{z^*})^*$

- In practice, a DFE suffers error propagation.

o Derivation of ZF-DFE from ZF-LE



$\hat{z}[n] = d[n] + \hat{N}_e[n]$
 where $\hat{N}_e[n]$ is a colored noise with PSD $\propto \frac{1}{Q(e^{j2\pi fT})}$



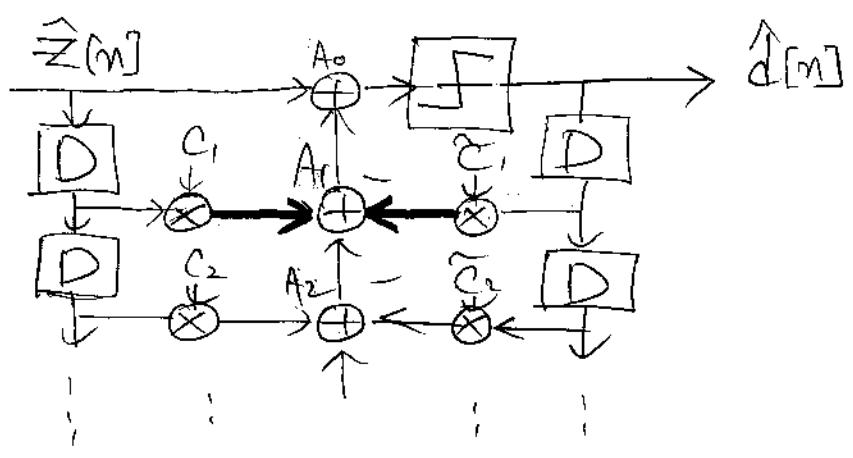
If the decision is perfect, then we may subtract $\hat{d}[n-1]$ from $\hat{z}[n-1]$ to obtain $\hat{N}_e[n-1]$.

Similarly, we may subtract $\hat{d}[n-2]$ from $\hat{z}[n-2]$ to obtain $\hat{N}_e[n-2]$, and so on.

- Since the noise sequence is colored, $\hat{N}_e[n-1], \hat{N}_e[n-2], \dots$ contain the information on $\hat{N}_e[n]$.

Thus, we may subtract the predicted value of $\hat{N}_e[n]$ from $\hat{z}[n]$ to have a better input to the slicer.

- Therefore, we want a system like

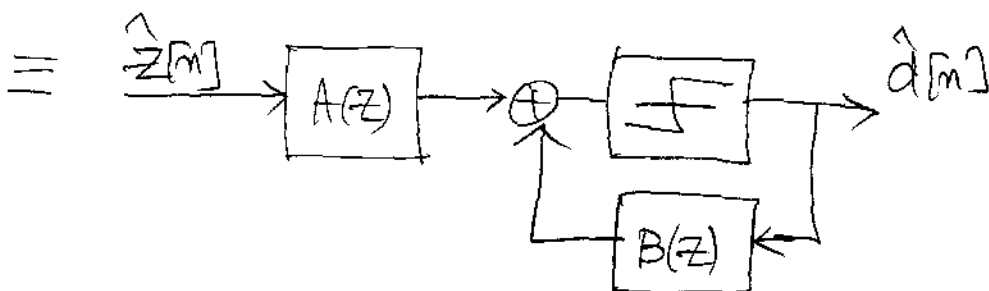
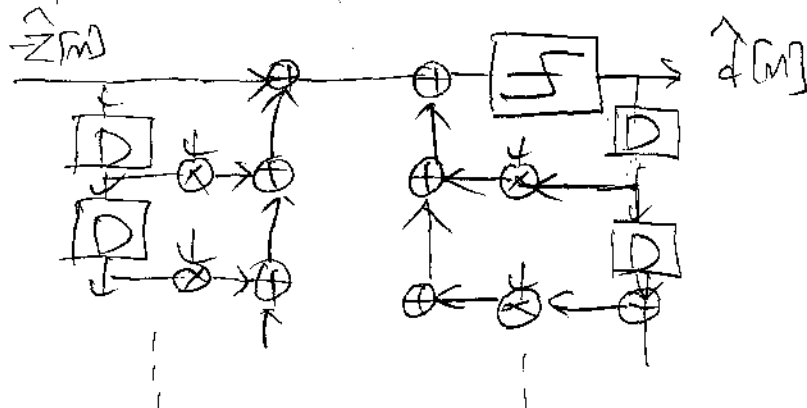


where the adder A_i does subtraction of $\tilde{c}_i \hat{d}[n-i]$ from $c_i \hat{z}[n-i] = c_i \hat{d}[n-i] + c_i \hat{N}_e[n-i]$ and

at the same time does subtraction of the best predicted value of $N[n]$ from $\hat{z}[n-i]$

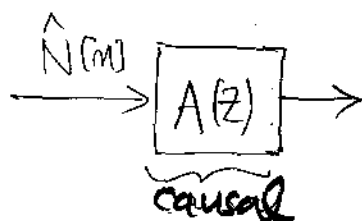
⑥

• This system is equivalent to



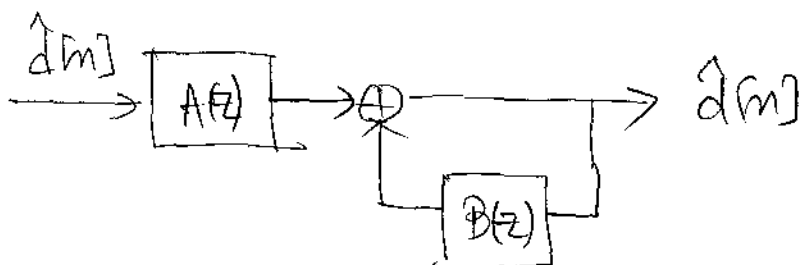
where

we want $A(z)$ to be the optimal causal predictor of $\hat{N}[m]$ when used as



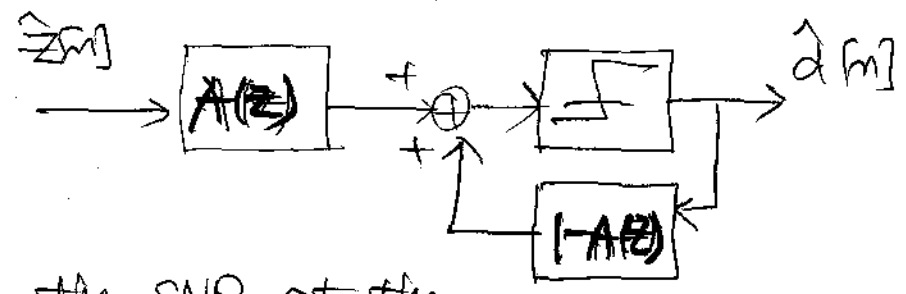
and

we want $B(z)$ to be the optimal canceller of the so-called postcursor ISI when used as



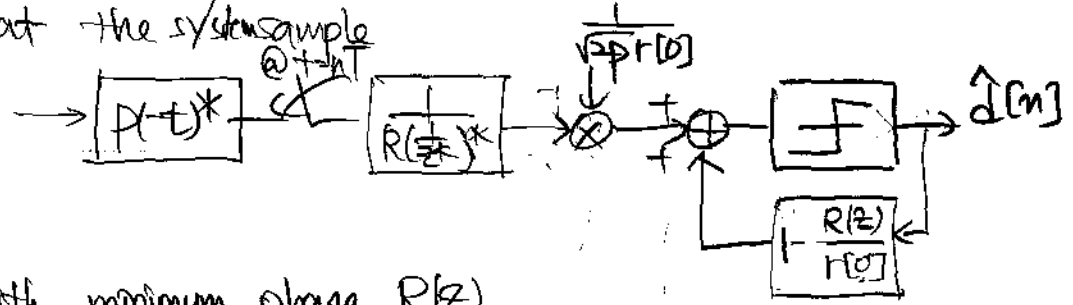
thus, $B(z) = 1 - \underbrace{A(z)}_{\text{causal}}$ and $A(z)$ is causal.

Therefore, we want a system



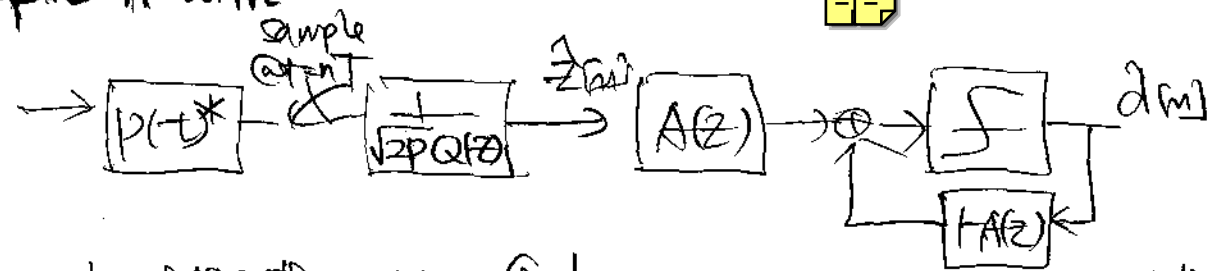
where the SNR at the slicer input is maximized.

We've already seen from the derivation of the ZF-DFE that the system sample



with minimum phase $R(z)$ maximizes the slicer input SNR and at the same time zero forces the input and whitens the noise input to the slicer.

Compare it with



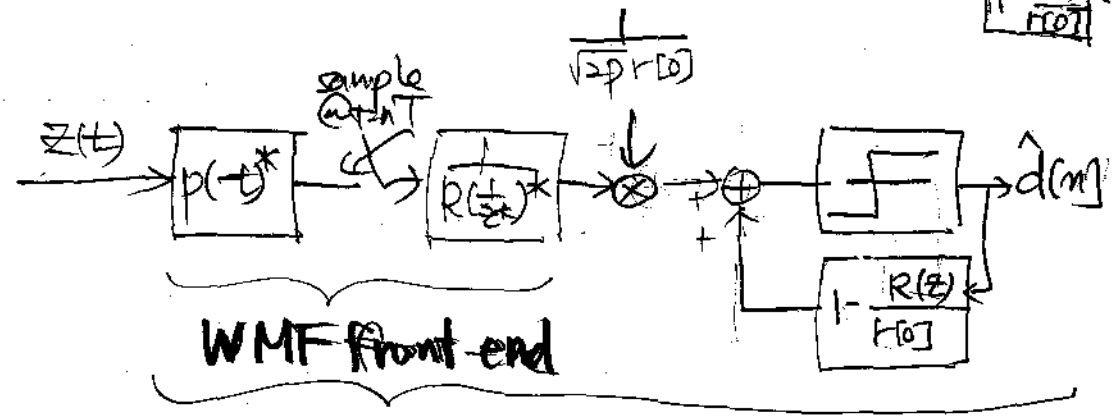
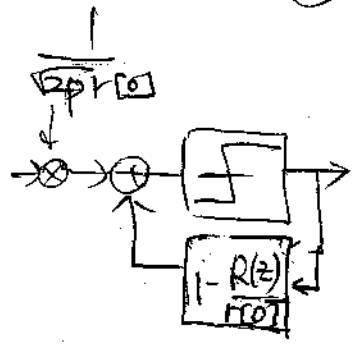
Then, by inspection we find $A(z) = \frac{R(z)}{r(z)}$ ← minic

← for this see LMMSE prediction, innovation is white

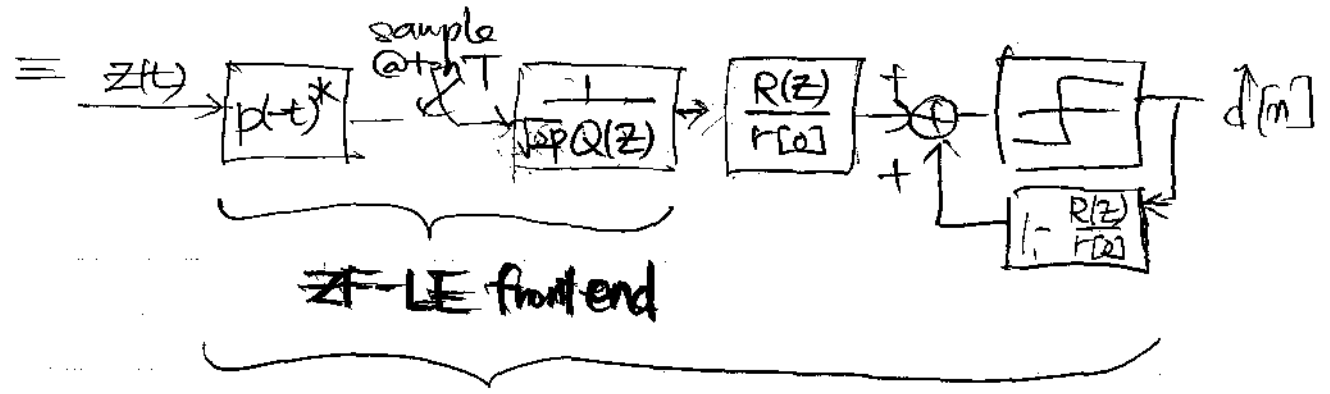
whitens the noise (i.e., optimally predicts the noise) and maximizes the input SNR to the slicer w/ ZF. (of course, p5 figure already shows $\hat{c}_i = -c_i - b_i$)

o Summary

- The WMF front end followed by
- does ZF-DFE



ZF-DFE



ZF-DFE

References

1. Barry, Lee, and Messerschmitt, Digital Communication, 3rd ed. pp. 355-365
2. Fischer, Precoding and signal shaping for Digital Transmission, pp. 49-71